

**DECT-Like System and Method Of Transceiving Information
Over The Industrial-Scientific-Medical Spectrum**

Cross Reference To Related Applications

5 This application is related to commonly assigned and co-pending US ⁵¹²
⁴¹⁷⁷⁴⁹³ patent application serial number 09/~~xxx,xxx~~ ^{Abandoned} (attorney's docket number P04658) ^{3/20/03}
entitled "System And Method For Concurrent Wireless Voice And Data
Communications" contemporaneously filed herewith and herein incorporated by
reference.

Background of the Invention

10 **1. Field of the Invention:**

The invention relates generally to wireless communications and more specifically to a system and method employing standard DECT hardware to operate in the Industrial-Scientific-Medical (ISM) band.

2. Description of Related Art:

15 The following background information is provided to aid in the understanding of the application of the present invention and is not meant to be limiting to the specific examples set forth herein. The so-called "Industrial-Scientific-Medical (ISM)" band allows for unlicensed wireless operation in the 2.4 GHz spectrum (as well as the 900MHz and 5.8 GHz spectrum) provided
20 however, that the power output is less than one watt and that some form of spread spectrum technology, i.e. Frequency Hopping Spread Spectrum (FHSS), Direct Sequence Spread Spectrum (DSSS) or a hybrid of FHSS and DSSS, is used to minimize interference. In the United States, 47 CFR Part 15 specifies the use of at least seventy-five (75) hopping frequencies between 2.4 GHz and 2.4835 GHz
25 and a minimum hop rate of 2.5 hops per second for FHSS systems. Several schemes have been proposed to use the ISM spectrum for wireless data

communication applications, such as for Wireless Local Area Networks (WLANs).

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5 In its 802.11 standard, the IEEE promulgated, inter alia, FHSS and DSSS definitions for the physical layer of a WLAN. For FHSS in North America and most of Europe, IEEE 802.11 requires 79 channels in 1 MHz steps beginning at 2.402 GHz and ending at 2.480 GHz with a minimum frequency hop of 6 MHz. Fig. 8 depicts the IEEE 802.11 protocol for packetizing information in a FHSS WLAN. One-hundred-twenty-eight (128) bits (a 96 bit preamble and 32 bit header) are sent to assist in synchronizing after a carrier hops from one
10 frequency to the next. Payload data then follows in sizes ranging from 1 to 4095 bytes.

Exhibit 100
15 An example of an ISM FHSS WLAN is the HomeCast™ Open Protocol (HOP™) from Alation Systems Inc. of Mountain View, California embodied in one form as the HomeFree™ Wireless Network product from Diamond Multimedia Systems, Inc. of Vancouver, Washington. The HOP system provides 79 channels and a maximum data throughput of 1 Mbps but employs a proprietary ISM baseband processor, requires a host processor to implement a software MAC, and does not support voice communications.

20 The so-called "PRISM I" chipset from the Intersil Corporation of Melbourne, Florida, is an example of a DSSS WLAN implementation in compliance with the IEEE 802.11 standard. The PRISM I chipset comprises six discrete integrated circuits, requires a proprietary baseband processor, and while maintaining IEEE 802.11 compliant, can only achieve a maximum data throughput of 2 Mbps and does not support voice communications. It can be
25 seen therefore, that the PRISM I DSSS solution while improving data throughput, also increases chip count and cost and locks the design into proprietary hardware.

By way of further background, reference is made to Fig. 1 that depicts the prior art Digital Enhanced Cordless Telecommunications (DECT) standard protocol promulgated by the European Telecommunications Standards Institute (ETSI). The DECT standard defines a Multiple Carrier, Time-Division-Multiple-
5 Access (TDMA), Time-Division-Duplex (TDD) protocol with ten channels (carrier frequencies) between 1881.792 MHz and 1897.344 MHz spaced 1.728 MHz apart. Each of the ten channels supports a ten-millisecond frame comprised of twenty-four time slots. TDD is provided by allocating twelve of the twenty-four slots for base station to cordless handset communications and
10 the other twelve slots for cordless handset to base station communications. Each time slot comprises 480 bits with a 32-bit preamble for synchronization, 388 bits for data and 60 bits for guard time. The 388 data bits are further divided into an A-field, a B-field and 4 parity bits for error detection. The A-field comprises an 8-bit header, 40 bits of control information and 16 cyclic redundancy check (CRC)
15 bits while the B-field provides 320 bits of data.

For speech applications, analog signals are digitized and encoded using adaptive differential pulse code modulation (ADPCM). Frequency hopping is employed to avoid interference by periodically assigning a different one of the ten channel frequencies to each of the twenty-four time slots. A form of
20 frequency shift keying known as Gaussian filtered, minimum shift keying (GMSK) is used to modulate the transmitted signal to provide continuous phase transitions between two adjacent symbols.

DECT enabled products are ubiquitous in Europe ranging from telephones and WLANs to cordless terminal mobility (CTM) applications
25 wherein a cordless handset operates with both private and public base stations. Unfortunately in the United States as well in other countries, the DECT spectrum between 1881.792 MHz and 1897.344 MHz is licensed for Personal Access

Communication Systems (PACS) and is not available for unlicensed applications such as WLANS.

From the foregoing it can be seen that there is a need for a system and method employing standard DECT hardware but operates in the unlicensed
5 Industrial-Scientific-Medical (ISM) spectrum.

Summary of the Invention

To overcome the limitations of the prior art described above, and to overcome other limitations that will become apparent upon reading and understanding the present specification, the present invention discloses a system
10 and method employing standard DECT hardware adapted for use in the Industrial-Scientific-Medical (ISM) Spectrum. A Frequency Hopping Spread Spectrum (FHSS), multiple carrier, time-division-multiple-access (TDMA), time-division-duplex (TDD) technique provides wireless communications over the ISM Spectrum while employing standard DECT hardware to map and morph the
15 DECT protocol to operate within the ISM spectrum. A baseband processor provides, among other things, slot and frame timing to a RF sub-module. The preferred, although not exclusive number of carrier frequencies is programmed to seventy-five ranging between 2401.122 MHz to 2479.813 MHz and spaced 1.063 MHz apart. Each of the seventy-five channels supports a ten-millisecond
20 frame preferably although not exclusively comprised of sixteen time slots. TDD is provided by allocating half of the slots for first to second tranceiving unit communications and the other half for second to first tranceiving unit communications.

A feature of the present invention is that standard hardware may be
25 employed for either DECT or ISM applications.

These and various other objects, features, and advantages of novelty which characterize the invention are pointed out with particularity in the claims

annexed hereto and forming a part hereof. However, for a better understanding of the invention, its advantages, and the objects obtained by its use, reference should be made to the drawings which form a further part hereof, and to the accompanying descriptive matter, in which there is illustrated and described a

5 specific example of DECT-Like System and Method of Transceiving Information Over The Industrial-Scientific-Medical Spectrum in accordance with the principles of the present invention.

Brief Description of the Drawings

Fig. 1 depicts a prior art diagram of the Digital Enhanced Cordless Telecommunications (DECT) standard protocol promulgated by the European Telecommunications Standards Institute (ETSI);

5 Fig. 2 depicts an illustrative but not limiting block diagram of a concurrent wireless voice and data communications system practiced in accordance with the principles of the present invention;

10 Fig. 3 depicts an illustrative but not limiting block diagram of a preferred Personal Access Device (PAD) practiced in accordance with the principles of the present invention;

15 Fig. 4 depicts a first exemplary but not limiting block diagram of a first preferred base station practiced in accordance with the principles of the present invention;

20 Fig. 5 depicts a second exemplary but not limiting block diagram of a second preferred base station practiced in accordance with the principles of the present invention;

25 Fig. 6 depicts an exemplary but not limiting block diagram of a preferred transceiver module practiced in accordance with the principles of the present invention;

30 Fig. 7 depicts the preferred protocol for a concurrent wireless voice and data communications system practiced in accordance with the principles of the present invention;

35 Fig. 8 depicts the preferred TDMA protocol for a concurrent wireless voice and data communications system practiced in accordance with the principles of the present invention, and,

40 Fig. 9 depicts a prior art IEEE 802.11 protocol for packetizing information in a frequency hopping spread spectrum wireless local area network.

Description of the Preferred Embodiment

The detailed description of the preferred embodiment for the present invention is organized as follows:

- 1.0 Exemplary System
- 5 2.0 Exemplary Personal Access Device (PAD)
- 3.0 Exemplary Base Station
- 4.0 Exemplary Transceiver Module
- 5.0 PAD to Base Station Synchronization
- 6.0 PAD-to-PAD Communications
- 10 7.0 Conclusion

This organizational table and the corresponding headings used in this detailed description are provided for the convenience of reference only and are not intended to limit the scope of the present invention. It is to be understood that while the preferred embodiment is described herein below with respect to

15 DECT and DECT-like wireless protocols, it has general applicability to any digital wireless communications technology. Certain terminology known to practitioners in the field of wireless communications is not discussed in detail in order not to obscure the disclosure. Moreover, in order not to obscure the disclosure with structural details which will be readily apparent to those skilled

20 in the art having the benefit of the description herein, the structure, control and arrangement of conventional circuits have been illustrated in the drawings by readily understandable block representations showing and describing details that are pertinent to the present invention. Thus, the block diagram illustrations in the figures do not necessarily represent the physical arrangement of the

25 exemplary system, but are primarily intended to illustrate the major structural components in a convenient functional grouping, wherein the present invention may be more readily understood.

Reference is now made to the accompanying drawings, which form a part hereof, and in which is shown by way of illustration specific embodiments in

30 which the invention may be practiced. It is to be understood that other

Semiconductor Corporation, Santa Clara, California, is coupled to DRAM 114 through an integrated DRAM controller (not shown) in the processor 112. A so-called "south bridge" chipset 116 is coupled to the processor 112, preferably through a PCI bus 113. The south bridge chipset 116 preferably includes an integrated ISA bus controller coupled to an ISA bus 115, a USB port 117 for supporting, inter alia, the keyboard 108 and FIFO buffers coupled to an audio CODEC 118. A flash ROM 111 is connected to the ISA bus 115 for storing code (such as an operating system and application programs) that is shadowed into DRAM 114 for execution by processor 112. The audio CODEC 118 converts digital signals to analog signals and drives speakers 121 and receives and converts analog signals from a monaural microphone 123 to digital signals for processing by processor 112. The display 120, which preferably is a DSTN or TFT LCD, is refreshed by a display adapter (not shown) that is integrated into either the processor 112 or chipset 116. The display 120 includes an overlaid programmable touch control panel 101 controlled by microcontroller 127 for use with removable stylus 106. The microcontroller 127 also provides charge profiling for rechargeable battery 129. The transceiver module 125 (discussed in more detail hereinbelow) is preferably, although not exclusively, connected to the ISA bus 115 for providing a wireless link to the base station 102.

20 3.0 Exemplary Base Station

Reference is now made to Fig. 4 that depicts a block diagram of the first preferred base station 102 without the transceiver module 125 installed, practiced in accordance with the principles of the present invention. While the first exemplary embodiment of the base station 102 is depicted as having a V.90 modem 135, those skilled in the art will readily recognize with the aid of the present disclosure, other forms of data network interfaces including but not limited to, ISDN, DSL and CATV modems and network adapters such as, but not limited to, Ethernet without departing from the scope of the present invention.

The V.90 modem interface 135 of the base station 102 is coupled to a Public switched telephone network (PSTN) through RJ11 jack 130. RJ11 jack 130 connects a first analog phone line through a first Digital Access Arrangement (DAA) 131 included within the V.90 modem 135. A combined CODEC/hybrid circuit 136 separates transmitted signals from received signals from the PSTN and converts the received signals into digital form. The received digital signals are operated on by a digital signal processor (DSP) 138 that executes code out of flash ROM 140 and SRAM 142 and 144 to provide, inter alia, interface control, AT command processing, and processing functions needed to perform signal modulations. Through execution of the code, the DSP 138 provides a command line AT interpreter, error checking, re-transmission, compression and decompression functions as well as necessary signal modulation/demodulation, adaptive filtering and encoding/decoding required for a V.90 standard modem.

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15 A second RJ11 jack 132 connects a second analog phone line from the PSTN to a second DAA 133 for voice reception/transmission. An optional third RJ11 jack 134 may be used to connect an external handset (not shown) to the base station 102. Optional LED indicators 143 controlled by DSP 138 display status of device ready, data and voice transmission in progress. Optional page key 145 may be provided to signal the transceiver module 150 (depicted in Fig. 5) through connector 149a to emit a page signal to the PAD 100. A power on reset (POR) circuit 147 provides reset signals to circuitry on the base station 102 as well as through connector 149a to the transceiver module 125, described in more detail herein below.

Reference is now made to Fig. 5 that depicts a block diagram of a second preferred base station 102' without the transceiver module 125 installed, practiced in accordance with the principles of the present invention. The second preferred base station 102' is constructed similar to that of the first base station 102 except for the elimination of secondary DAA 133 and the addition of the

relay 137. In the second preferred version of the base station 102', relay 137, which is controlled via the baseband processor 180 in transceiver module 125, switches the PSTN coupled through RJ11 jack 130 and DAA 131 to either the data network adapter (e.g. modem) or the ancillary analog voice channel (provided by the baseband processor 180 in transceiver module 125), all of which is discussed in more detail herein below.

Although while only one phone line is connected to the base station 102', the user of a PAD 100 can utilize the data network (via modem) and still be made aware of an incoming call via the transceiver module 125 that provides call notification and caller ID to allow the user of the PAD 100 to switch from the data network to the voice network. For example, this may manifest itself through a pop-up window on the PAD 100 notifying a single phone line user of PAD 100 (who may be surfing the world-wide-web) of an incoming phone call thus permitting the user of PAD 100 to switch from surfing the web to answer the phone call.

4.0 Exemplary Transceiver Module

Reference is now made to Fig. 6 that depicts by way of illustration an exemplary but not limiting block diagram of the preferred transceiver module 125 practiced in accordance with the principles of the present invention. The transceiver module 125 comprises an antenna 152 (multiple antennas for diversity), an RF sub-module 150 coupled to a baseband processor 180, a flash ROM 182 and RAM 183 to store code for execution by the baseband processor 180 and a mating connector 149b for connecting to either the base station connector 149a or to the ISA bus 115 in the PAD 100.

The RF sub-module 150 includes, inter alia, a band pass filter (BPF) 154 coupled to a transmit/receive switch 156. Received data from the transmit/receive switch 156 is conditioned by a low noise amplifier (LNA) 158

and a BPF 160 prior to being sent to a mixer within single chip radio transceiver 162. Transmitted data from the single chip radio transceiver 162 is passed through a LNA 164 and transmit power amplifier 166 prior to being sent to transmit/receive switch 156. An exemplary but not limiting example of a single
5 chip containing the BPFs 154 and 160, LNAs 158 and 164, transmit/receive switch 156 and power amplifier 166 is the AU2404T RF front-end integrated circuit from Alation Systems Inc. of Mountain View, California. Those skilled in the art, with the aid of the present disclosure, will recognize other forms and solutions for elements 154, 156, 158, 160, 164 and 166 without departing from the
10 spirit and scope of the present invention.

The single chip radio transceiver 162 in combination with BPFs 168 and 170 and voltage controlled oscillator (VCO) 172 and loop filter 174 down convert (receive) or up convert (transmit) data to/from baseband processor 180. The preferred although not exclusive embodiment for the single chip radio
15 transceiver 162 is the LMX3162 transceiver from National Semiconductor Corporation of Santa Clara, CA, described in the National Analog and Interface Products Databook (and accompanying CD-ROM), 1999, which is herein incorporated by reference. The RF sub-module 150 is available from ALPS Electric Co, Ltd. of Tokyo, Japan under the model numbers UGSA4-402A
20 (without antenna diversity) and UGSA4-502A (with antenna diversity) for 2.4 GHz operation and under the model numbers UGSE2-402A (without antenna diversity) and UGSE2-502A (with antenna diversity) for 1.8 GHz (DECT) operation.

The baseband processor 180 preferably comprises a CODEC and at least
25 one sub-processor that executes code stored in flash ROM 182 and RAM 183 to handle, inter alia, audio, signal and data processing for tone generation, echo canceling and to program slot and frame timing for the RF sub-module 150. In general, the code executed by the baseband processor 180 in the transceiver

module 125 is preferably layered in adherence with the Open Systems Interconnection (OSI) model, the details of which are known to one skilled in the art. The preferred although not exclusive embodiment for the baseband processor 180 is the SC14424 baseband processor from National Semiconductor Corporation of Santa Clara, CA, described in detail in Appendix A hereto.

Reference is now made to Fig. 7 that depicts the preferred protocol for a concurrent wireless voice and data communications system practiced in accordance with the principles of the present invention. The preferred protocol is a multiple carrier, Time-division-multiple-access (TDMA), Time-division-duplex (TDD) system. The preferred programmable, although not exclusive number of carrier frequencies is seventy-five ranging between 2401.122 MHz to 2479.813 MHz and spaced 1.063 MHz apart. Those skilled in the art having the benefit of the description herein will appreciate other numbers of carrier frequencies (e.g. ten), frequency spectrums (e.g. 1881.792 MHz to 1897.344 MHz) and spacings (e.g. 1.728 MHz apart) without departing from the scope the present invention. Each of the seventy-five channels supports a ten-millisecond frame preferably comprised of sixteen time slots. Those skilled in the art having the benefit of the description herein will appreciate other numbers of time slots without departing from the scope the present invention. Symmetrical TDD is provided by allocating half (i.e. eight of the sixteen slots) for base station to PAD communications and the other half (i.e. eight slots) for PAD to base station communications. Asymmetrical TDD is contemplated as well wherein base station to PAD communications consume more slots (e.g. twelve slots) than PAD to base station communications (i.e. four slots) or vice versa. Those skilled in the art having the benefit of the description herein will appreciate other asymmetric numbers of slot allocations for base station to PAD communications and vice versa without departing from the scope the present invention.

Each time slot preferably comprises a 32-bit preamble for synchronization, a 64 bit A-field for signaling and a B-field comprising 320 bits and 4 bits for CRC. Each of the sixteen time slots receives/transmits on one of the seventy-five carrier channels that preferably changes in a pseudo-random fashion, to one of the other seventy-four carrier channels after two consecutive frames thus providing fifty (50) hops/second. Those skilled in the art having the benefit of the description herein will appreciate other number of frequency carriers, hopping patterns and frequency hop periods without departing from the scope the present invention.

Reference is now made to Fig. 8 that depicts the preferred TDMA protocol in more detail. In the preferred embodiment, symmetric TDMA (with respect to both voice/data and base station/PAD communications) is provided by allocating time slots 1, 2, 3 and 9, 10, 11 for data communication between base station and PAD and PAD and base station, respectively, and time slots 4, 5, 6 and 12, 13, 14 for voice communication between base station and PAD and PAD and base station, respectively. Time slots 7 and 15 are reserved, time slot 8 is allocated to program the transmit carrier frequency in the single chip radio transceiver 162 and slot 16 is allocated to program the receive carrier frequency.

Asymmetrical TDMA is contemplated as well wherein data communications consume more slots (e.g. twelve slots) than voice communications (i.e. four slots) or vice versa. As mentioned above, asymmetry with respect to base station/PAD communications is contemplated and it is further contemplated that asymmetric base station/PAD communications may be used in combination with asymmetric data/voice TDMA. Those skilled in the art having the benefit of the description herein will appreciate other asymmetric numbers of slot allocations for base station/PAD communications and/or data/voice communications without departing from the scope the present invention.

To achieve frequency hopping, the transmit and receive carrier frequencies are changed by the baseband processor 180 reprogramming a phase locked loop (PLL) in the single chip radio transceiver 162. The transmit and receive carrier frequencies are changed by the baseband processor 180 in a pseudo-random fashion, to one of the other seventy-four carrier channels after two consecutive frames thus providing fifty (50) hops/second.

Referring to Fig. 8b, a time slot dedicated to data allocates 80 bits in the B field to a Forward Error Correction Code (FECC). The remaining 240 bits are payload data for processing by the PAD 100. A time slot dedicated to voice allocates the entire 320 bits in the B field to voice information since voice is tolerant to dropouts in bit patterns.

So-called "multi-slot" operation (e.g. double slot) is further contemplated wherein adjacent slots share a single set of sync, signaling, CRC bits and optionally, FECC bits. In a single data slot, the sync, signaling, CRC and FECC bits consume 180 out of the 420 bits allocated to a slot. By way of illustration and not of limitation, a double data slot shares one 32-bit preamble for synchronization, one 64 bit A-field for signaling, one set of 80 FECC bits and one set of four CRC bits, thus providing 660 payload data bits over two slots instead of the standard 480 bits – a 37.5% increase in bandwidth. Those skilled in the art having the benefit of the description herein will appreciate other multi-slot configurations (e.g. quad slots) and allocation of overhead bits without departing from the scope the present invention.

5.0 PAD to Base Station Synchronization

On power up, the baseband processor 180 in the transceiver module 125 of the PAD 100 executes code to set the received carrier frequency to a reference channel and to scan for incoming data during a time period of two frames (e.g. 20 milliseconds). If an A-Field with a correct CRC is detected, the baseband

processor 180 continues to sample and decode A-Fields every frame (e.g. 10 milliseconds) for the expected ID of the base station 102. If a timeout occurs, the baseband processor 180 code restarts with the next carrier frequency channel. If the correct ID for the base station 102 is received, the A-field decoding continues

5 until the current slot number, frame number, multi-frame number, and carrier channel in which the base station 102 is scanning are received. The base band processor 180 updates its corresponding internal variables and enters into a locked state when this information is received.

The baseband processor 180 in the transceiver module 125 of the base

10 station 102 executes code to fix the transmit carrier frequency to a reference channel until the PAD 100 synchronizes. Thereafter, the baseband processor 180 executes code to change the transmit carrier frequency in the transceiver module 125 of the base station 102 every two frames (e.g. 20 milliseconds) in a pseudo-random sequence so long as correct A-Fields are found in two consecutive

15 frames. The PAD 100 continues to receive A-Fields until the expected ID of the base station 102 is received or a timeout occurs. If a timeout occurs, the process is restarted with the reference channel frequency.

6.0 PAD-to-PAD Communications

PAD-to-PAD communications is further contemplated wherein multiple

20 PADs communicate with one another through a common base station 102. By way of illustration and not of limitation, a second PAD 100' is added to Fig. 2 wherein PADs 100 and 100' communicate with one another via the base station 102. Each PAD 100 and base station have a unique ID associated with it that can be embedded in the A-field (64 bits of signaling) of the intended target. The

25 baseband processor 180 in the transceiving module 125 of the base station 102 detects whether the received ID in the A-field is intended for the base station 102. If so, the voice/data information associated with that A-field is processed and routed to the respective voice/data network tethered to the base station 102. If

not, the base station 102 relays the voice/data information associated with that A-field onto the intended PAD 100'.

7.0 Conclusion

Although the Detailed Description of the invention has been directed to
5 certain exemplary embodiments, various modifications of these embodiments, as well as alternative embodiments, will be suggested to those skilled in the art. The invention encompasses any modifications or alternative embodiments that fall within the scope of the Claims.